TS3002

A 1V/1µA Easy-to-Use Silicon Oscillator/Timer

FEATURES

- Ultra Low Supply Current: 1µA at 25kHz
- Supply Voltage Operation: 0.9V to 1.8V
- Programmable Frequency Range:
  - o 5.2kHz ≤ FOUT ≤ 90kHz (BOOST = GND)
  - o 5.2kHz ≤ FOUT ≤ 290kHz (BOOST = VDD)
- FOUT Period Drift: 0.044%/°C
- PWMOUT Duty Cycle Range: 12% to 90%
- Single Resistor and Capacitor Set Output Frequency
- Output Driver Resistance: 160Ω

APPLICATIONS

Portable and Battery-Powered Equipment
Low-Parts-Count Nanopower Oscillator
Compact Nanopower Replacement for Crystal and Ceramic Oscillators
Nanopower Pulse-width Modulation Control
Nanopower Pulse-position Modulation Control
Nanopower Clock Generation
Nanopower Sequential Timing

DESCRIPTION

The TS3002 is the industry's first and only single-supply CMOS oscillator fully specified to operate at 1V while consuming a 1µA supply current at an output frequency of 25kHz. This oscillator is compact, easy-to-use, and versatile. Optimized for ultra-long life, battery-powered applications, the TS3002 is the first oscillator in the "NanoWatt Analog™" high-performance analog integrated circuits portfolio. The TS3002 can operate from single-supply voltages from 0.9V to 1.8V.

Requiring only a resistor and a capacitor to set the output frequency, the TS3002 represents a 66% reduction in pcb area and a factor-of-10 reduction in power consumption over other CMOS-based integrated circuit oscillators. When compared against industry-standard 555-timer-based products, the TS3002 offers up to 93% reduction in pcb area and four orders of magnitude lower power consumption.

The TS3002 is fully specified over the -40°C to +85°C temperature range and is available in a low-profile, 8-pin 2x2mm TDFN package with an exposed back-side paddle.

TYPICAL APPLICATION CIRCUIT

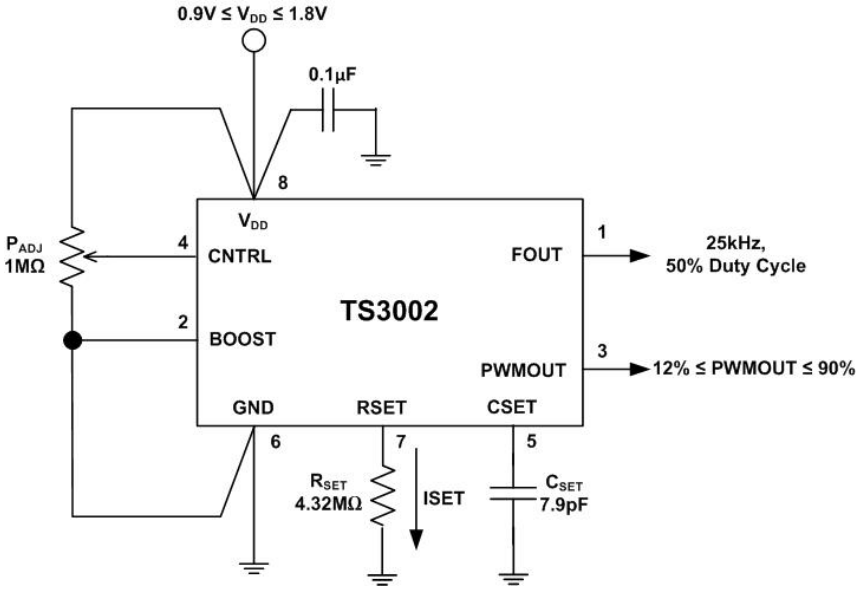

Table 1: FOUT vs R_{SET}, C_{SET} = 7.9pF

R_{SET} (MΩ)	FOUT (kHz)
1	106
2.49	43
4.32	25
6.81	16
9.76	11

Table 2: FOUT vs C_{SET}, R_{SET} = 4.32MΩ

C_{SET} (pF)	FOUT (kHz)
5	39
7.9	25
10	19
15	13
20	10

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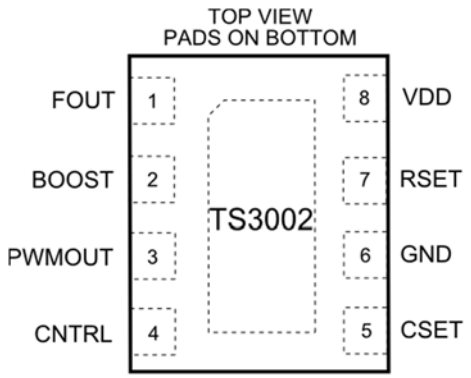
ABSOLUTE MAXIMUM RATINGS

V_{DD} to GND ..... -0.3V to +2V
V_{CNTRL} to GND ..... -0.3V to +2V
RSET to GND ..... -0.3V to +2V
CSET to GND ..... -0.3V to +2V
FOUT, PWMOUT to GND ..... -0.3V to +2V
Short Circuit Duration FOUT, PWMOUT to GND or V_{DD}
..... Continuous

Continuous Power Dissipation (T_A = +70°C)
8-Pin TDFN (Derate at 23.8mW/°C above +70°C) 1951mW
Operating Temperature Range ..... -40°C to +85°C
Storage Temperature Range ..... -65°C to +150°C
Lead Temperature (Soldering, 10s) ..... +300°C

Electrical and thermal stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other condition beyond those indicated in the operational sections of the specifications is not implied. Exposure to any absolute maximum rating conditions for extended periods may affect device reliability and lifetime.

PACKAGE/ORDERING INFORMATION



2mm x 2mm x 0.75mm DFN-EP 8L
TD822 Package

ORDER NUMBER	PART MARKING	CARRIER	QUANTITY
TS3002ITD822	AAH	Tape & Reel	-----
TS3002ITD822T		Tape & Reel	3000

Lead-free Program: Silicon Labs supplies only lead-free packaging.

Consult Silicon Labs for products specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

V_{DD} = 1V, V_{CNTRL} = V_{DD}, V_{BOOST} = 0V, R_{SET} = 4.32MΩ, C_{SET} = 7.9pF, $R_{LOAD(FOUT)}$ = Open Circuit, $C_{LOAD(FOUT)}$ = 0pF, $C_{LOAD(PWM)}$ = 0pF unless otherwise noted. Values are at T_A = 25°C unless otherwise noted. See Note 1.

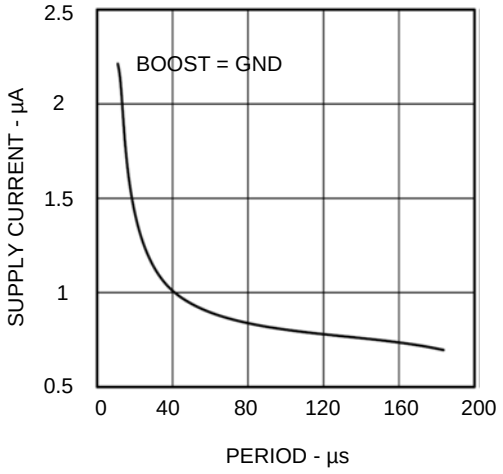
PARAMETER	SYMBOL	CONDITIONS			MIN	TYP	MAX	UNITS
Supply Voltage	V_{DD}				0.9	1	1.8	V
Supply Current	I_{DD}					1	1.5	µA
				-40°C ≤ T_A ≤ 85°C			2.8	
		V_{CNTRL} = 0.15 x V_{DD}				2.1	3.7	
				-40°C ≤ T_A ≤ 85°C			5.4	
		V_{BOOST} = V_{DD}				2.16	3.2	
				-40°C ≤ T_A ≤ 85°C			4.8	
		V_{BOOST} = V_{DD}, V_{CNTRL} = 0.15 x V_{DD}				3.6	5.3	
				-40°C ≤ T_A ≤ 85°C			7.3	
FOUT Period	t_{FOUT}				37	40.6	44	µs
				-40°C ≤ T_A ≤ 85°C	34.7		45.6	
		V_{BOOST} = V_{DD}			36	39.5	43	
				-40°C ≤ T_A ≤ 85°C	33		48	
FOUT Period Line Regulation	$\Delta t_{FOUT}/V$	1V ≤ V_{DD} ≤ 1.8V				1.3		%/V
			V_{BOOST} = V_{DD}			-1.6		
FOUT Period Temperature Coefficient	$\Delta t_{FOUT}/\Delta T$					0.044		%/°C
		V_{BOOST} = V_{DD}				0.086		
PWMOUT Duty Cycle	DC(PWMOUT)	V_{CNTRL} = 0.03 x V_{DD}			4.5	8.9	13	%
		V_{CNTRL} = 0.15 x V_{DD}			44	49.3	54	
		V_{CNTRL} = 0.27 x V_{DD}			83	90.5	97	
		V_{CNTRL} = 0.03 x V_{DD}	V_{BOOST} = V_{DD}		4.5	8.5	12.5	
		V_{CNTRL} = 0.15 x V_{DD}			47	50.4	54	
		V_{CNTRL} = 0.27 x V_{DD}			86	91.2	96	
FOUT, PWMOUT Rise Time	t_{RISE}	See Note 2, C_L = 15pF				8.6		ns
FOUT, PWMOUT Fall Time	t_{FALL}	See Note 2, C_L = 15pF				7.9		ns
FOUT Jitter		See Note 3				0.08		%
RSET Pin Voltage	V(RSET)					0.3		V
CNTRL Output Current	I_{CNTRL}					25	45	nA
				-40°C ≤ T_A ≤ 85°C			80	
PWMOUT Enable	V_{PWM_EN}	(V_{DD} - V_{CNTRL}), 0.9V < V_{DD} < 1.8V			375			mV
PWMOUT Disable	V_{PWM_DIS}	(V_{DD} - V_{CNTRL}), 0.9V < V_{DD} < 1.8V					131	mV
BOOST Enable	V_{IH}	(V_{DD} – V_{BOOST}), 0.9V < V_{DD} < 1.8V					77	mV
BOOST Disable	V_{IL}	0.9V < V_{DD} < 1.8V					77	mV
BOOST Input Current	I_{BOOST}					10		nA
High Level Output Voltage, FOUT and PWMOUT	V_{DD} - V_{OH}	I_{OH} = 1mA				160		mV
Low-level Output Voltage, FOUT and PWMOUT	V_{OL}	I_{OL} = 1mA				140		mV

Note 1: All devices are 100% production tested at T_A = +25°C and are guaranteed by characterization for T_A = T_{MIN} to T_{MAX}, as specified.
Note 2: Output rise and fall times are measured between the 10% and 90% of the V_{DD} power-supply voltage levels. The specification is based on lab bench characterization and is not tested in production.
Note 3: Timing jitter is the ratio of the peak-to-peak variation of the period to the mean of the period. The specification is based on lab bench characterization and is not tested in production.

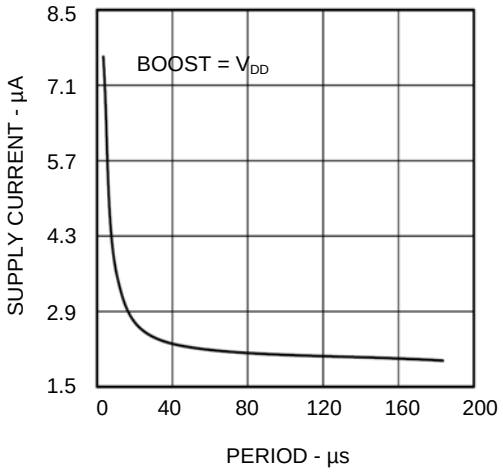
TYPICAL PERFORMANCE CHARACTERISTICS

V_{DD} = 1V, V_{CNTRL} = V_{DD}, V_{BOOST} = 0V, R_{SET} = 4.32MΩ, C_{SET} = 7.9pF, $R_{LOAD(FOUT)}$ = Open Circuit, $C_{LOAD(FOUT)}$ = 5pF, unless otherwise noted.
Values are at T_A = 25°C unless otherwise noted.

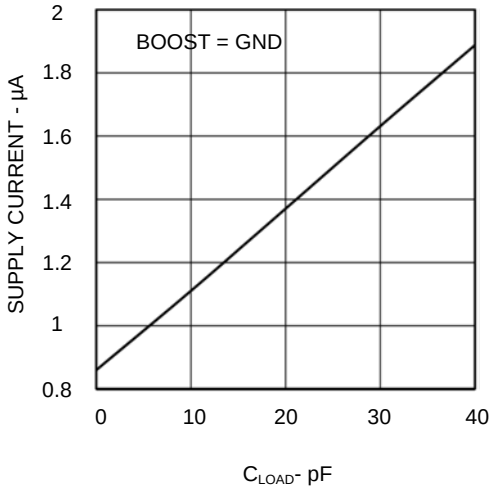
Supply Current vs FOUT Period



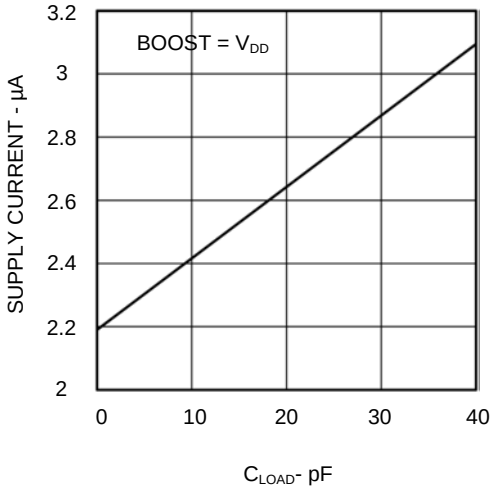
Supply Current vs FOUT Period



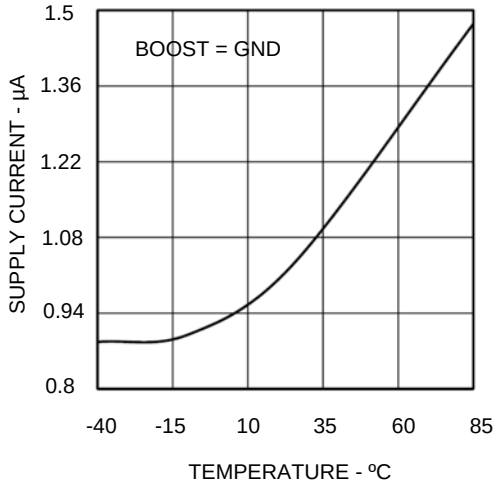
Supply Current vs $C_{LOAD(FOUT)}$



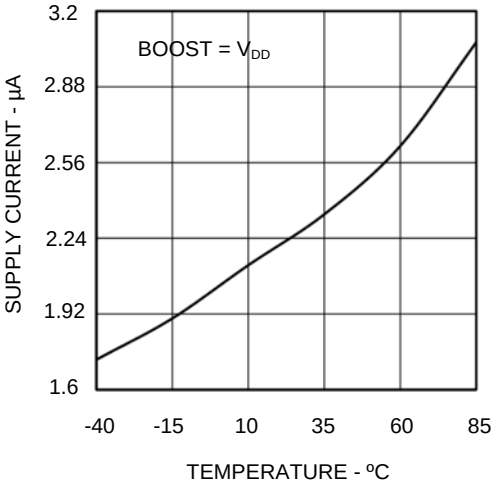
Supply Current vs $C_{LOAD(FOUT)}$



Supply Current vs Temperature



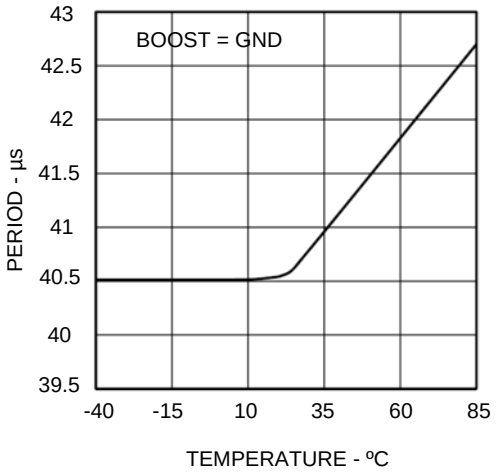
Supply Current vs Temperature



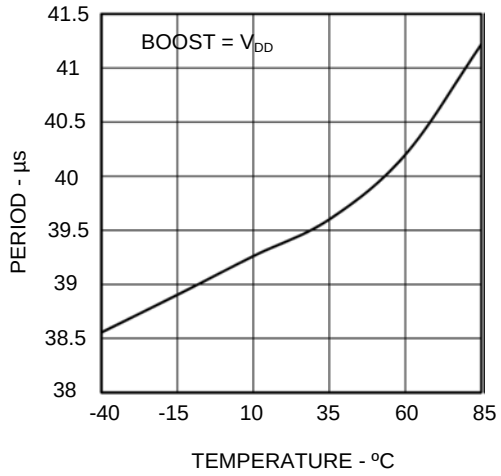
TYPICAL PERFORMANCE CHARACTERISTICS

V_{DD} = 1V, V_{CNTRL} = V_{DD}, V_{BOOST} = 0V, R_{SET} = 4.32MΩ, C_{SET} = 7.9pF, $R_{LOAD(FOUT)}$ = Open Circuit, $C_{LOAD(FOUT)}$ = 5pF, unless otherwise noted. Values are at T_A = 25°C unless otherwise noted.

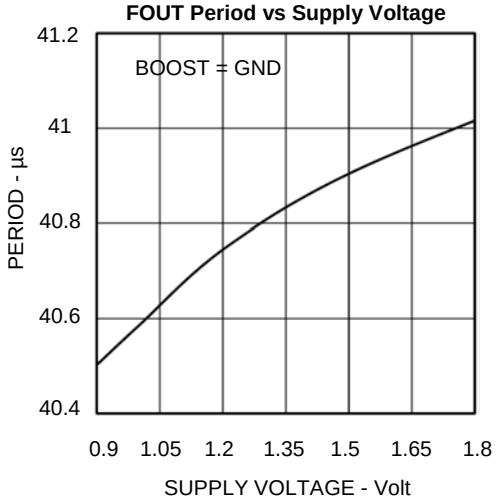
FOUT Period vs Temperature



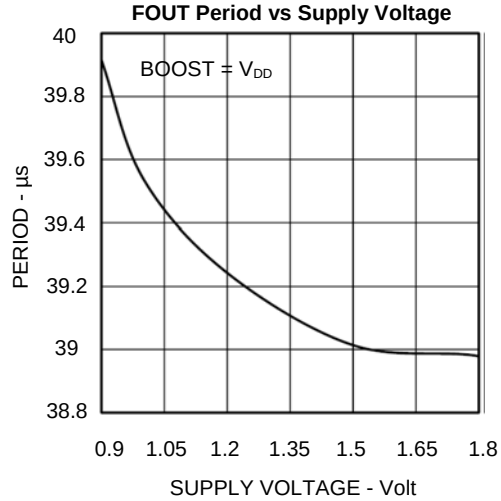
FOUT Period vs Temperature



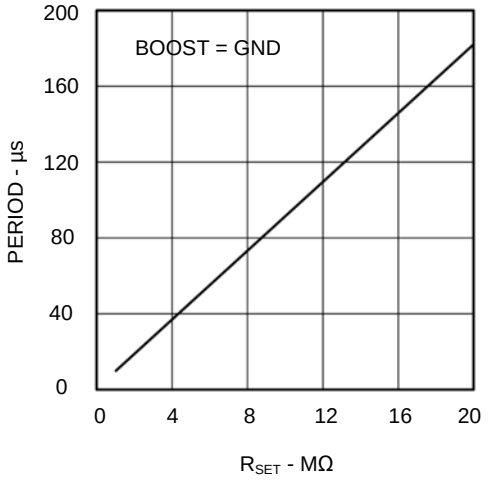
FOUT Period vs Supply Voltage



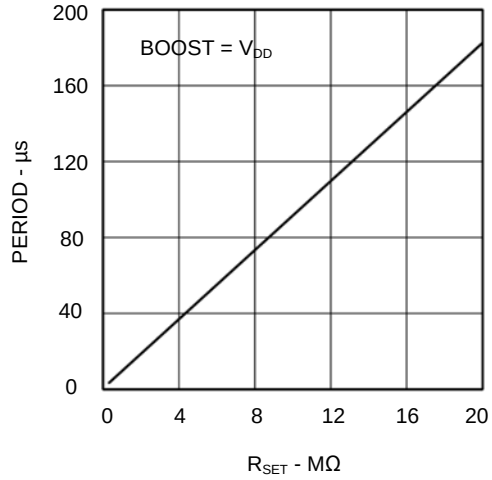
FOUT Period vs Supply Voltage



Period vs R_{SET}



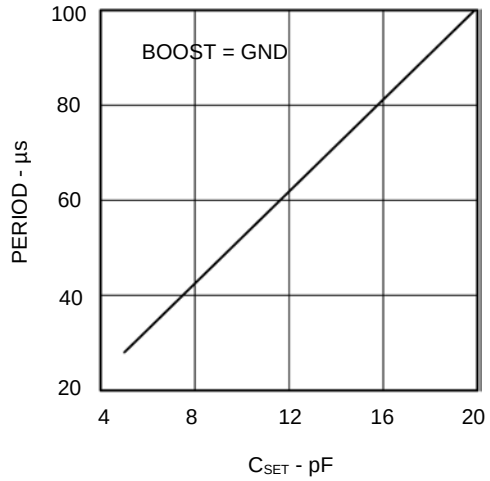
Period vs R_{SET}



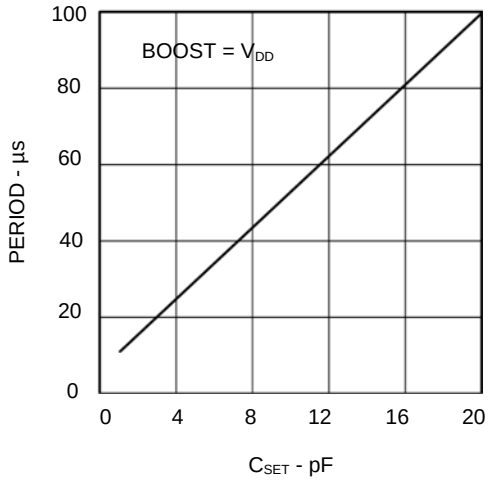
TYPICAL PERFORMANCE CHARACTERISTICS

V_{DD} = 1V, V_{CNTRL} = V_{DD}, V_{BOOST} = 0V, R_{SET} = 4.32MΩ, C_{SET} = 7.9pF, $R_{LOAD(FOUT)}$ = Open Circuit, $C_{LOAD(FOUT)}$ = 5pF, unless otherwise noted.
Values are at T_A = 25°C unless otherwise noted.

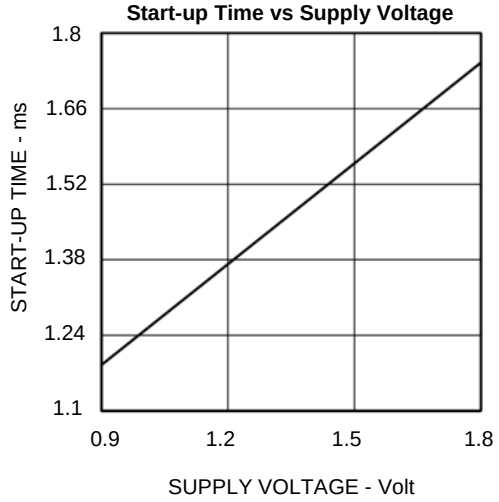
Period vs C_{SET}



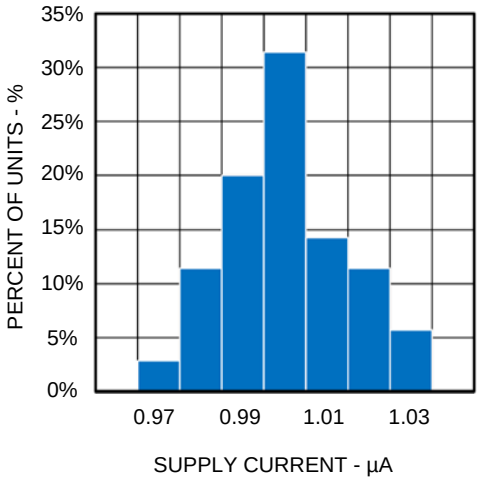
Period vs C_{SET}



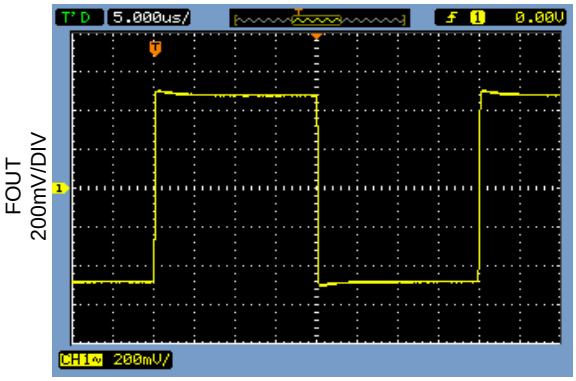
Start-up Time vs Supply Voltage



Supply Current Distribution

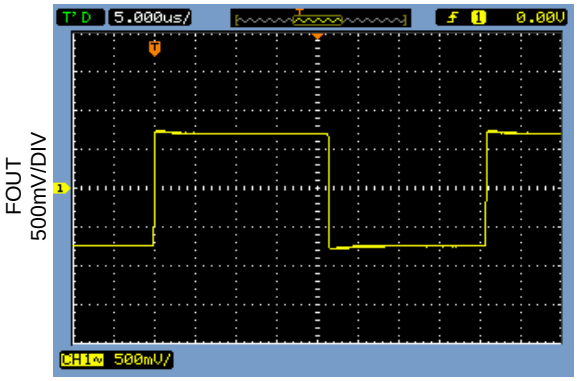


FOUT Transient Response
V_{DD} = 1V, BOOST = V_{DD}, C_{LOAD} = 47pF



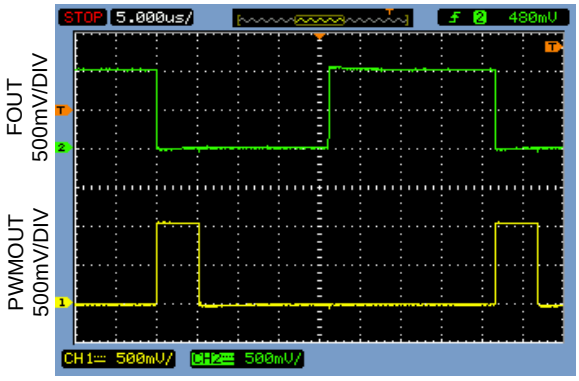
5µs/DIV

FOUT Transient Response
V_{DD} = 1.5V, BOOST = V_{DD}, C_{LOAD} = 47pF



5µs/DIV

FOUT and PWMOUT Transient Response
V_{DD} = 1V, V_{CNTRL} = 0.035 x V_{DD}, BOOST = V_{DD}, C_{LOAD} = 22pF

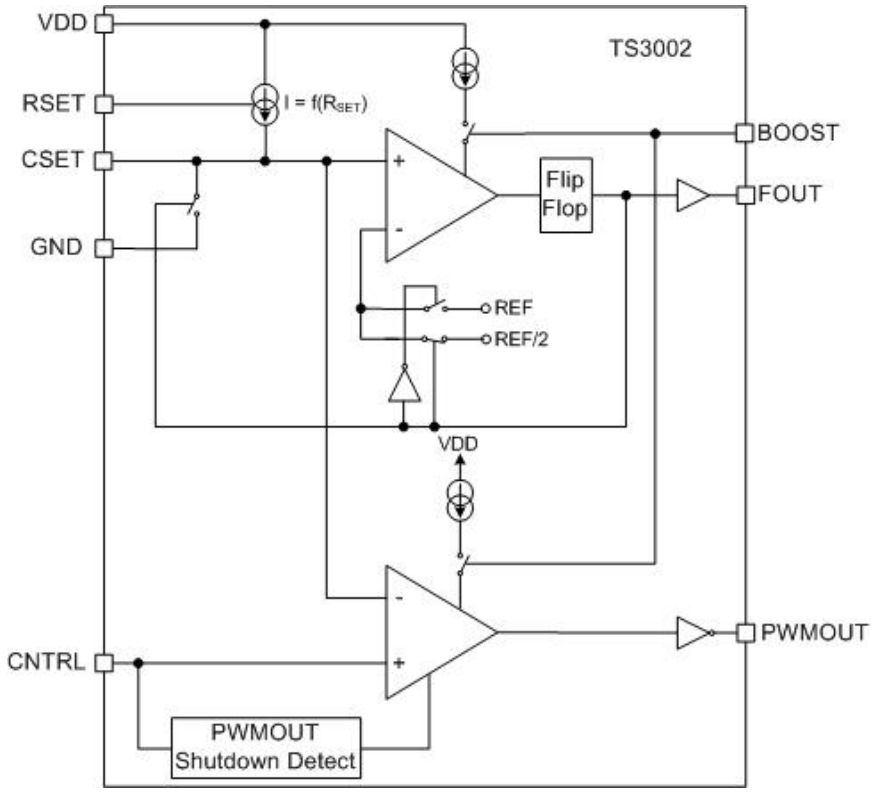


5µs/DIV

PIN FUNCTIONS

PIN	NAME	FUNCTION
1	FOUT	Fixed Frequency Output. A push-pull output stage with an output resistance of 160Ω, the FOUT pin swings from GND to V_{DD}. For lowest power operation, capacitive loads should be minimized and resistive loads should be maximized.
2	BOOST	BOOST Input. A digital switch input, BOOST controls the propagation delay of the primary timing comparator in the TS3002's master oscillator subcircuit. Connecting the BOOST pin to GND sets the maximum programmable oscillator frequency to ~90kHz.Connecting the BOOST pin to VDD reduces the comparator's propagation delay and increases the maximum programmable master oscillator's frequency to 290kHz.
3	PWMOUT	Pulse-width Modulated Output. A push-pull output stage with an output resistance of 160Ω, the PWMOUT pin is wired anti-phase with respect to FOUT and swings from GND to V_{DD}. For lowest power operation, capacitive loads should be minimized and resistive loads should be maximized.
4	CNTRL	PWMOUT Enable and Duty Cycle Control Input. An analog input pin, the V_{CNTRL} pin voltage enables the TS3002's PWM engine and controls the duty cycle at PWMOUT from 12% (V_{CNTRL} = 0.03 x V_{DD}) to 90% (V_{CNTRL} = 0.27 x V_{DD}). Enabling the PWM engine increases the TS3002's nominal operating supply current. To disable the TS3002's PWM engine, CNTRL shall be connected to V_{DD}.
5	CSET	FOUT Programming Capacitor Input. A 7.9pF capacitor connected from this pin to GND in junction with a 4.32MΩ resistor at the RSET pin sets the TS3002's internal oscillator's output period to ~40µs (25kHz). The maximum capacitance value is 22pF.
6	GND	Ground – Connect this pin to the system's analog ground plane.
7	RSET	FOUT Programming Resistor Input. A 4.32MΩ resistor connected from this pin to GND sets the TS3002's internal oscillator's output period to 40µs (25kHz). For optimal performance, the composition of the RSET resistor shall be consistent with tolerances of 1% or lower. The RSET pin voltage is 0.3V at a 1V supply.
8	VDD	Power Supply Voltage Input. While the TS3002 is fully specified at 1V, the supply voltage range is 0.9V ≤ V_{DD} ≤ 1.8V. It is always considered good engineering practice to bypass the V_{DD} pin with a 0.1µF ceramic decoupling capacitor in close proximity to the TS3002.
EP	-----	Exposed paddle is electrically connected to GND.

BLOCK DIAGRAM



THEORY OF OPERATION

The TS3002 is a user-programmable oscillator where the period of the square wave at its FOUT terminal is generated by an external resistor and capacitor pair. The output frequency is given by:

$$\text{FOUT (kHz)} = \frac{1}{t_{FOUT}\ (\mu s)} = \frac{1E6}{k \cdot R_{SET}(M\Omega) \times C_{SET}(pF)}$$

where the scalar k is approximately 1.19. With an R_{SET} = 4.32MΩ and a C_{SET} = 7.9pF, the output frequency is approximately 25kHz with a 50% duty cycle. As design aids, Tables 1 lists TS3002's typical FOUT for various standard values for R_{SET} with C_{SET} = 7.9pF and Table 2 lists typical FOUT for various standard values for C_{SET} with R_{SET} = 4.32MΩ.

The TS3002 also provides a separate PWM output

Table 1: FOUT vs R_{SET}, C_{SET} = 7.9pF

R_{SET} (MΩ)	FOUT (kHz)
1	106
2.49	43
4.32	25
6.81	16
9.76	11

Table 2: FOUT vs C_{SET}, R_{SET} = 4.32MΩ

C_{SET} (pF)	FOUT (kHz)
5	39
7.9	25
10	19
15	13
20	10

signal at its PWMOUT terminal that is anti-phase with respect to FOUT. In addition, applying a voltage at the CNTRL both enables the TS3002's internal PWM engine as well as adjusting the duty cycle from 12% to 90%. A dc control voltage equal to 0.03 x VDD applied to the CNTRL pin enables the PWM engine to set the duty cycle to 12%. A dc control voltage equal to 0.27 x VDD increases the duty cycle to 90% and connecting CNTRL to VDD disables the PWM engine altogether. Configured for nominal operation (PWM engine OFF, BOOST pin to GND), the supply current of the TS3002 is 1µA; enabling the PWM engine increases the TS3002 operating supply current as shown in the electrical specification table. The BOOST pin controls the propagation delay of the TS3002's internal comparators. When BOOST is connected to GND, the TS3002's maximum programmable operating frequency is ~90kHz. Connecting the BOOST pin to VDD reduces the propagation delay of the internal oscillators, thereby extending the high end maximum operating frequency to 290kHz.

APPLICATIONS INFORMATION

Minimizing Power Consumption

To keep the TS3002's power consumption low, resistive loads at the FOUT and PWMOUT terminals increase dc power consumption and therefore should be as large as possible. Capacitive loads at the FOUT and PWMOUT terminals increase the TS3002's transient power consumption and, as well, should be as small as possible.

One challenge to minimizing the TS3002's transient power consumption is the probe capacitance of oscilloscopes and frequency counter instruments. Most instruments exhibit an input capacitance of 15pF or more. Unless buffered, the increase in transient load current can be as much as 400nA.

To minimize capacitive loading, the technique shown in Figure 1 can be used. In this circuit, the principle of series-connected capacitors can be used to reduce the effective capacitive load at the TS3002's FOUT and PWMOUT terminals.

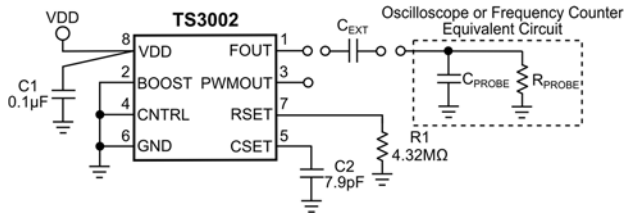


Figure 1: Using an External Capacitor in Series with Probes Reduces Effective Capacitive Load.

To determine the optimal value for C_{EXT} once the probe capacitance is known by simply solving for C_{EXT} using the following expression:

$$C_{EXT} = \frac{1}{\frac{1}{C_{LOAD(EFF)}} - \frac{1}{C_{PROBE}}}$$

For example, if the instrument's input probe capacitance is 15pF and the desired effective load capacitance at either or both FOUT and PWMOUT terminals is to be ≤5pF, then the value of C_{EXT} should be ≤7.5pF.

TS3002 Start-up Time

As the TS3002 is powered up, its FOUT terminal (and PWMOUT terminal, if enabled) is active once the applied VDD is higher than 0.9 volt. Once the applied VDD is higher than 0.9 volt, the master oscillator achieves steady-state operation within 1.2ms.

Current- and Voltage-Controlled Oscillators

The TS3002 can be configured into a Current-Controlled Oscillator as shown in Figure 2.

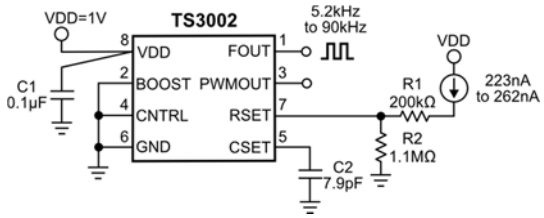


Figure 2: Configuring the TS3002 into a Current-Controlled Oscillator.

With a current source sourcing a current of 223nA to 262nA, FOUT can generate an output signal with a frequency range of 5.2kHz to 90kHz. In a similar manner, a Voltage-Controlled Oscillator can be configured as shown in Figure 3. In this case, a voltage source sourcing a voltage of 290mV to

341mV can generate an FOUT output signal frequency range of 5.2kHz to 90kHz as well. It is recommended to use resistor values with a 1% tolerance.

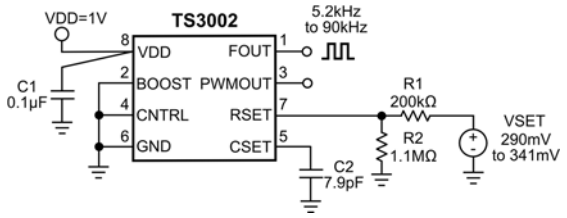


Figure 3: Configuring the TS3002 into a Voltage-Controlled Oscillator.

Using a Potentiometer to Trim the TS3002's Output Frequency

By using a fixed resistor and a potentiometer, the output frequency of the TS3002 can be trimmed as shown in Figure 4. By selecting a fixed resistor R1 with a tolerance of 0.1% and a potentiometer P1 with a 5% tolerance, the output frequency can be trimmed to provide a ±2% trimming range. As shown in Figure 5, R1+P1 and C2 set the output frequency to 25.052kHz when P1 = 0Ω and with P1 =200kΩ, the resulting output frequency is 24.024kHz.

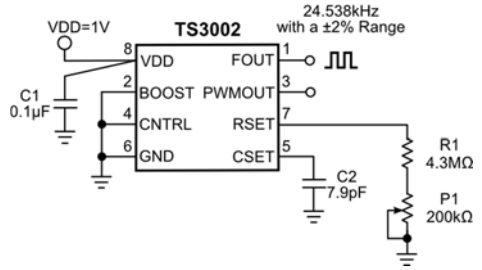


Figure 4: Using a Fixed Resistor and a Potentiometer to Trim the TS3002's Output Frequency.

Using Standard Resistors to Increase FOUT Resolution

The TS3002 can be configured to provide a 0.1% resolution on the output frequency as shown in Figure 5. To do so, R1 can be set to approximately 10% of the value selected for R2. In addition, R2 and R1 should be chosen with a 0.1% and 1% tolerance, respectively. Since R2 is 90% of the total resistance, it has the largest impact on the resolution of the output frequency. With R1 = 91kΩ and R2 = 910kΩ, the output frequency is 90kHz and with R1 = 400kΩ and R2 = 4MΩ, the output frequency is 23kHz.

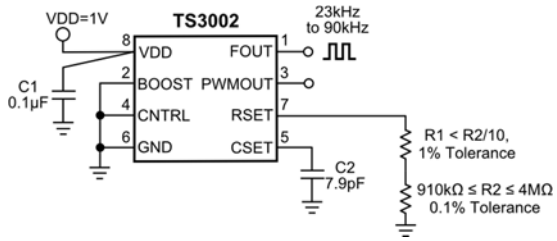
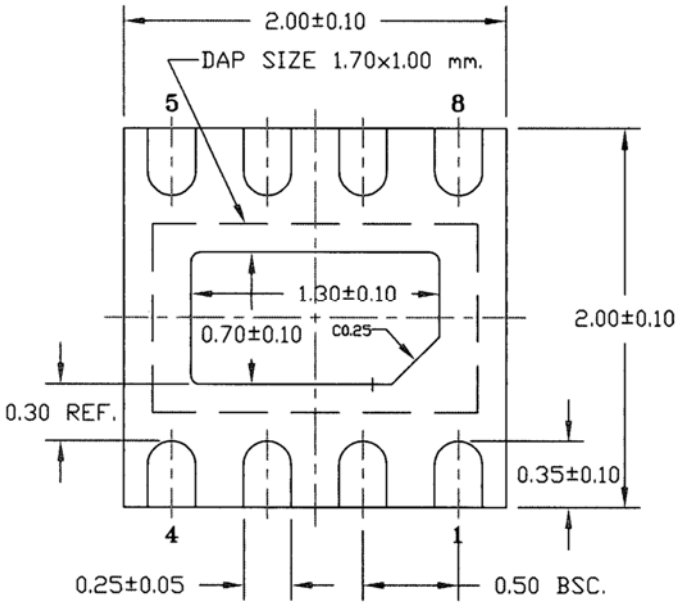


Figure 5: Setting the TS3002's Output Frequency to 0.1% Resolution using Standard Resistors.

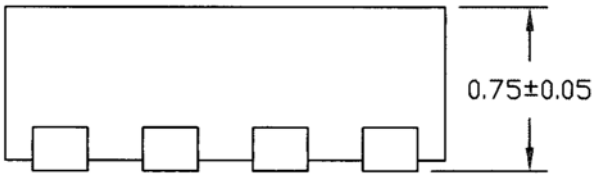
PACKAGE OUTLINE DRAWING

8-Pin TDFN22 Package Outline Drawing

(N.B., Drawing not to scale; all dimensions in mm; JEDEC MO-229 compliant)



BOTTOM VIEW



SIDE VIEW

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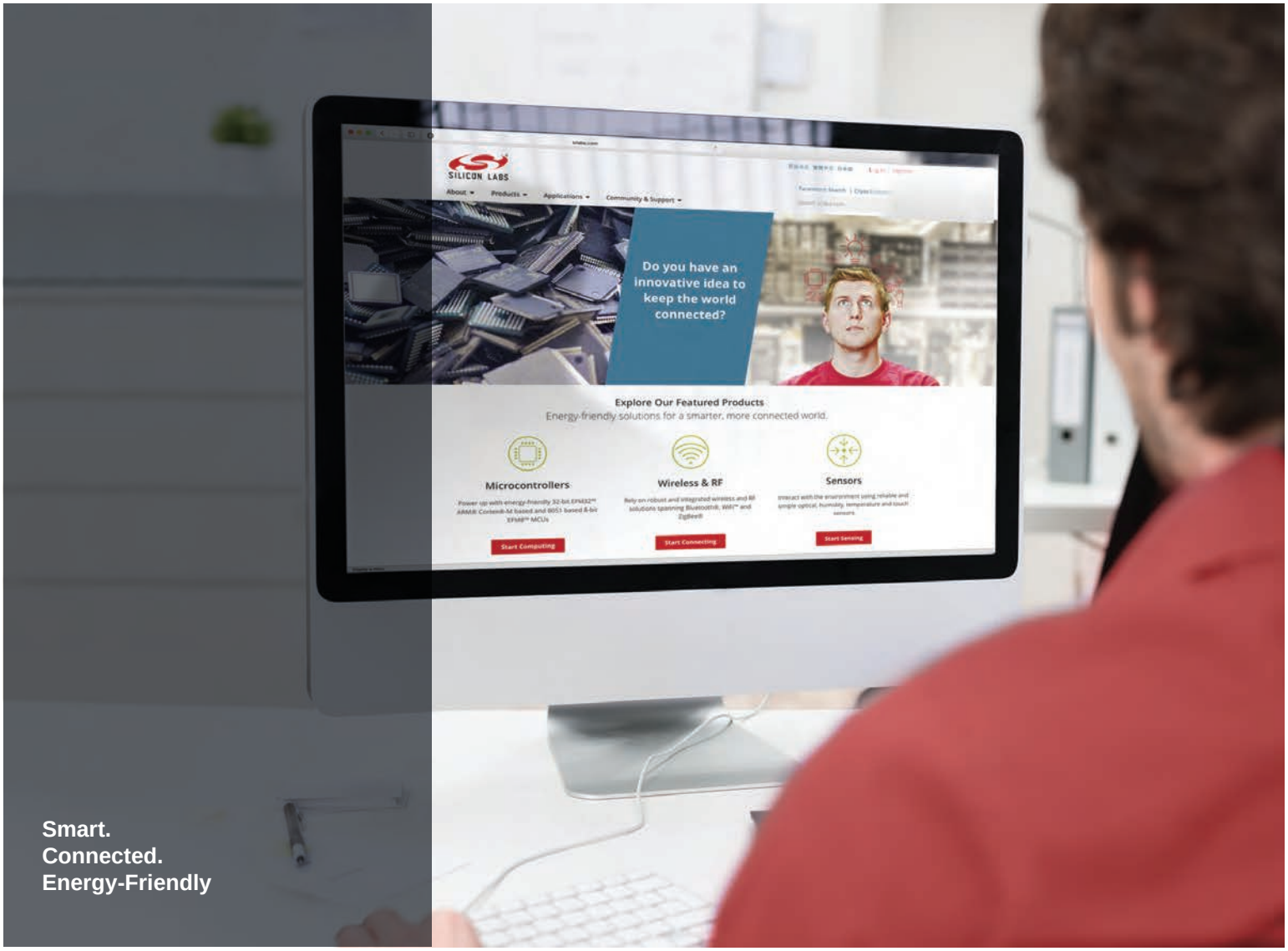

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